

# Forward message passing detector for probe storage.

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Joint work with:

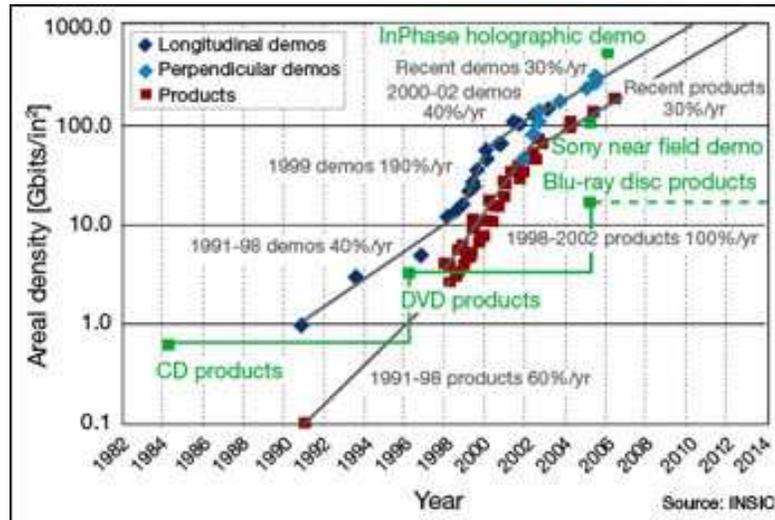
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## Plan.

- Thermomechanical probe storage.
- Detection/decoding schemes for probe storage devices.
- Channel model for probe storage.
- Soft output detection: forward message passing (FMP).
- FMP detector for probe storage.
- Performance analysis: mutual information and BER
- Performance analysis: SER
- Conclusions

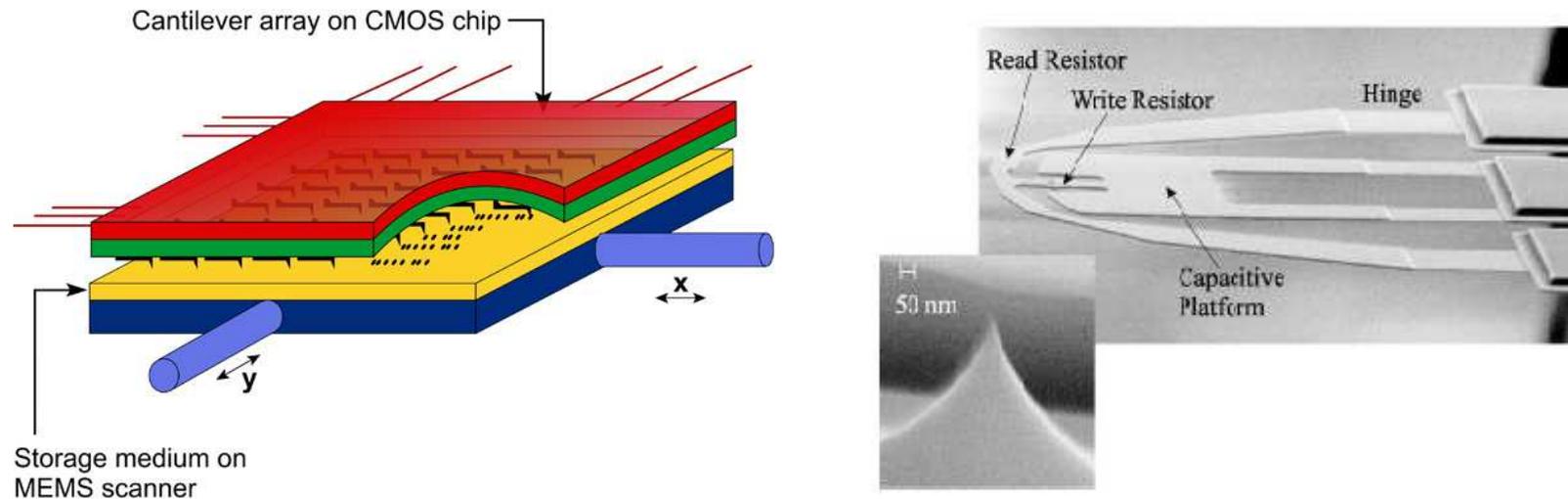
# Probe storage.

# Storage density trends.



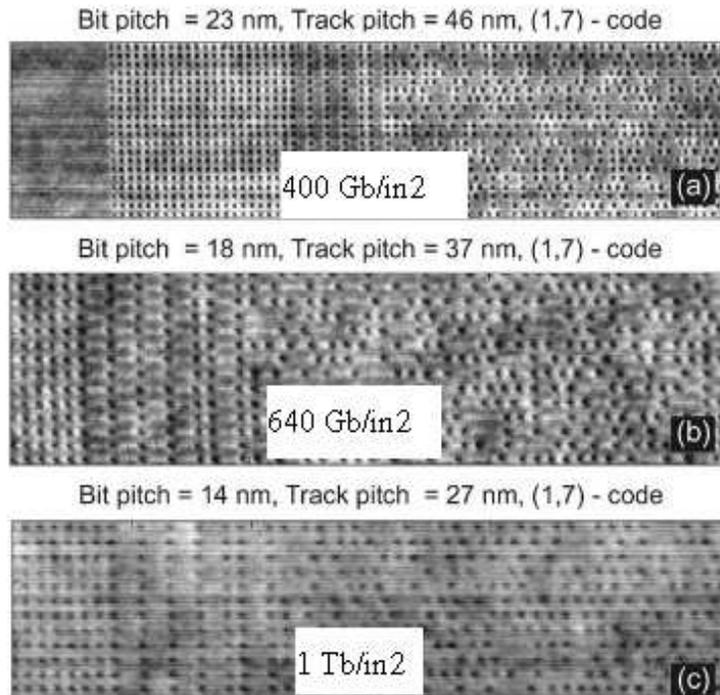
- Magnetic storage density is expected to reach  $1 \text{ Tb}/\text{in}^2$
- Further growth is limited by the superparamagnetic effect.
- Probe storage has already demonstrated  $1 \text{ Tb}/\text{in}^2$  with  $4 \text{ Tb}/\text{in}^2$  demonstrators being developed.

# IBM's Thermomechanical Probe Storage Concept



- Thin polymer medium is positioned under the array of  $64 \times 64$  atomic force probes.
- Each probe operates in its own field of size  $100\mu m \times 100\mu m$ . Tip radius  $\sim 10\text{ nm}$ .
- Encoded data are stored as pits on the surface of the medium.

## Read/Write



- Writing: the probe's tip is heated and pressed into the softened polymer film
  - Reading: The probe heated to a smaller  $T$  follows the landscape of the polymer surface
- A probe inserted into a pit is cooler than the probe whose tip touches the surface. These variations are captured using a thermo-resistive sensor

# Channel Model

## Non-linear Inter-symbol Interference (ISI)

$$I_k(x_{k-1}, x_k, x_{k+1})$$

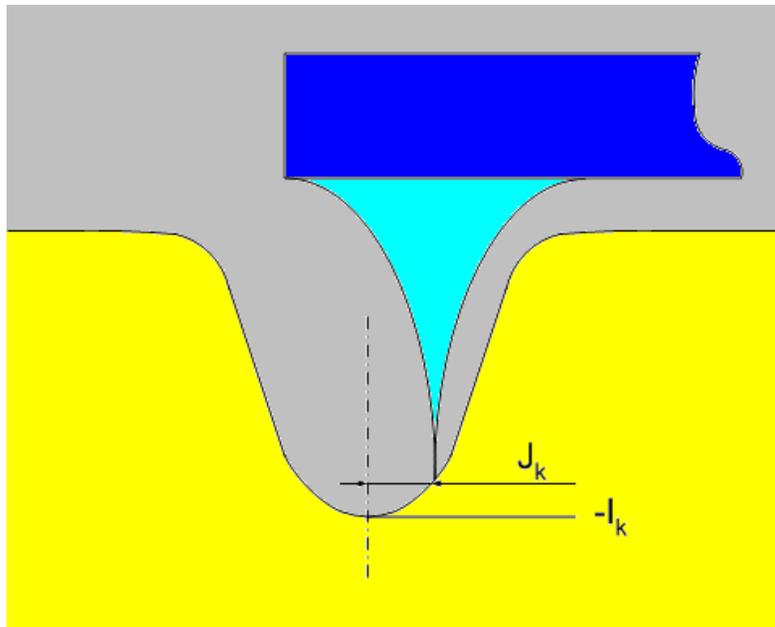
$$= x_k$$

$$+(\alpha - 1)x_k x_{k+1}$$

$$+\beta x_k x_{k-1}$$

- Ideal readout at the  $k$ -th sampling point
- A signal due to an isolated pit at  $k$
- Reduction in the signal strength due to plastic displaced from the  $(k + 1)$ -st pit
- Signal enhancement due to plastic displaced into the  $(k - 1)$ -st pit
- For experiments at  $1 Tb/in^2$ ,  $\alpha \approx 0.8$ ,  $\beta \approx 0.1$
- $\alpha$ ,  $\beta$  depend on write parameters, tip shape and medium material properties

## Position Jitter



- Jitter = positioning error
- $J \ll \text{pit width}$
- $\Delta I_k \sim J_k^2$
- Over 40% of total noise power is due to data-dependent position jitter

$$r_k \approx I_k \cdot \left( 1 - \left( \frac{\sigma_j}{h} W_k \right)^2 \right) + \sigma_e N_k,$$

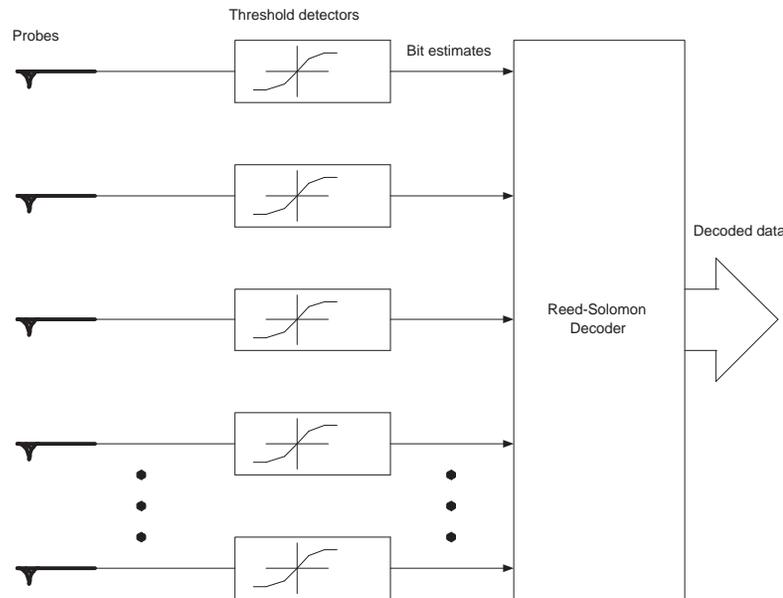
$W_k, N_k$  are independent normal r. v.'s,  $h$  is the pit's radius of curvature;  $\sigma_j, \sigma_e$  are the strengths of jitter and electronics noise correspondingly

## Statistics of Signal Distortion

- Let  $\eta_k = \frac{r_k - I_k}{I_k}$
- Let  $\epsilon = \frac{\sigma_j}{h}$ ,  $\delta = \frac{\sigma_e}{I_k}$ .
- $\rho(\eta) \sim \frac{\text{Const}_-}{|\eta|^{1/2}} e^{2\frac{\eta}{\epsilon^2}}$ ,  $\eta \ll -\epsilon^2$
- $\rho(\eta) \sim \frac{\text{Const}_+}{\eta^{1/2}} e^{-\frac{\eta^2}{2\delta^2}}$ ,  $\eta \gg \delta$
- **Signal distortion is non-Gaussian**

# Detection/decoding

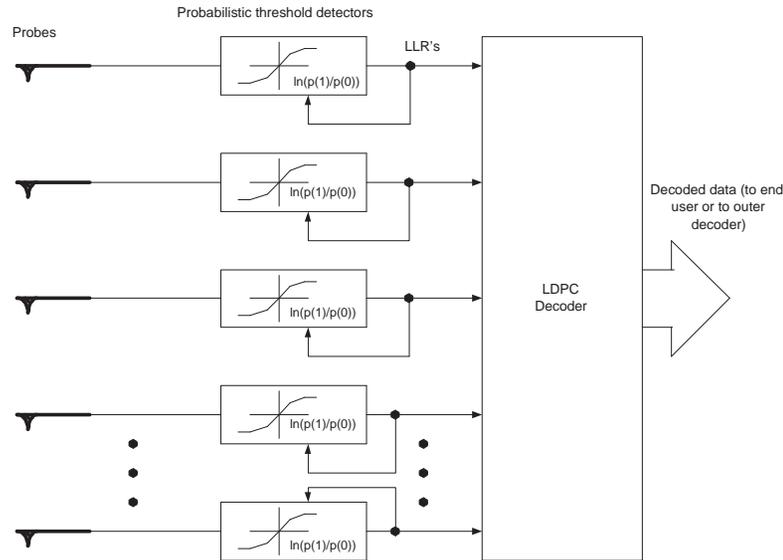
## The currently employed scheme



- Read channel: hard output threshold detector
- ECC: Reed-Solomon code

- HDD read channel: a sector of data is detected as the *most likely* binary string given the digitised received string using Viterbi algorithm.
- A significant increase of recording density beyond  $1 \text{ Tb}/\text{in}^2$  would require a significantly more advanced detection decoding scheme

## Desired scheme



- Read channel: soft output data detector
- ECC: Soft input decoder for  $LDPC$ ,  $LDPC \circ RS$ ,  $SPC \circ RS$ , etc. code

- MAP detector per probe is too complex
- Any easy ways to generate soft outputs?

# Soft detection via forward message passing

## Soft threshold detector.

$$LLR_k \stackrel{def}{=} \ln \frac{Pr(x_k=1|r_k)}{Pr(x_k=0|r_k)} \stackrel{Bayes}{=} \ln \frac{Pr(r_k|x_k=1)}{Pr(r_k|x_k=0)}$$

- Threshold bit estimate:  $\hat{x}_k = \text{sign} LLR_k$
- Information contained in  $r_{k'}: k' \neq k$  is not used in the computation of  $LLR_k$ .

## Forward message passing detector.

- Assume that  $r_k$  depends on  $x_k, x_{k\pm 1}$  only.
- Let  $LLR_k = \ln \left( \frac{\Pr(x_k=1|\vec{r}_{k+1})}{\Pr(x_k=0|\vec{r}_{k+1})} \right)$ , where  $\vec{r}_k = \dots r_{k-3}r_{k-2}r_{k-1}r_k$ . Then

$$\Pr(\vec{r}_k \mid x_{k+1}, x_k, x_{k-1}) = \frac{1}{2} \Pr(r_k \mid x_{k+1}, x_k, x_{k-1}) \sum_{x_{k-2}=0}^1 \Pr(\vec{r}_{k-1} \mid x_k, x_{k-1}, x_{k-2})$$

- Message is an 8-dimensional vector of probabilities  $\Pr(\vec{r}_k \mid x_{k+1}, x_k, x_{k-1})$  propagated left-to-right using transfer matrix built out of conditional probabilities  $\Pr(r_k \mid x_{k+1}, x_k, x_{k-1})$ .

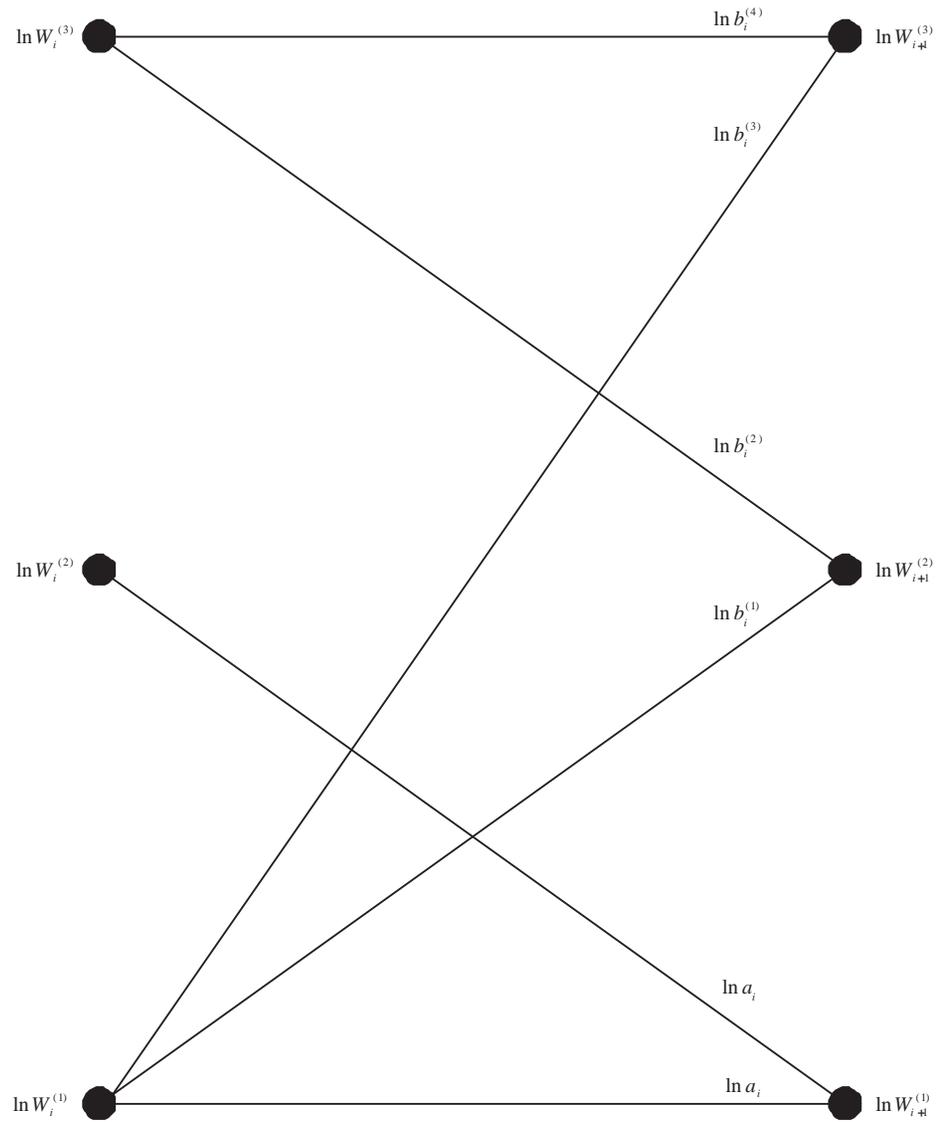
## Transition matrix

$$\mathbf{T} = \begin{pmatrix} \alpha_i & \alpha_i & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & \alpha_i & \alpha_i & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \beta_i^{(1)} & \beta_i^{(1)} & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & \beta_i^{(2)} & \beta_i^{(2)} \\ \alpha_i & \alpha_i & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & \alpha_i & \alpha_i & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \beta_i^{(3)} & \beta_i^{(3)} & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & \beta_i^{(4)} & \beta_i^{(4)}, \end{pmatrix}$$

where  $\alpha$ 's and  $\beta$ 's are conditional probabilities.

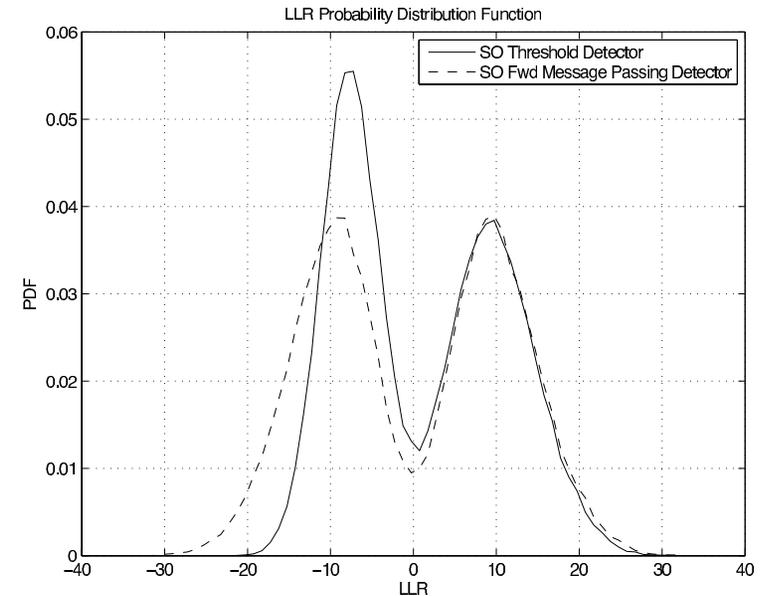
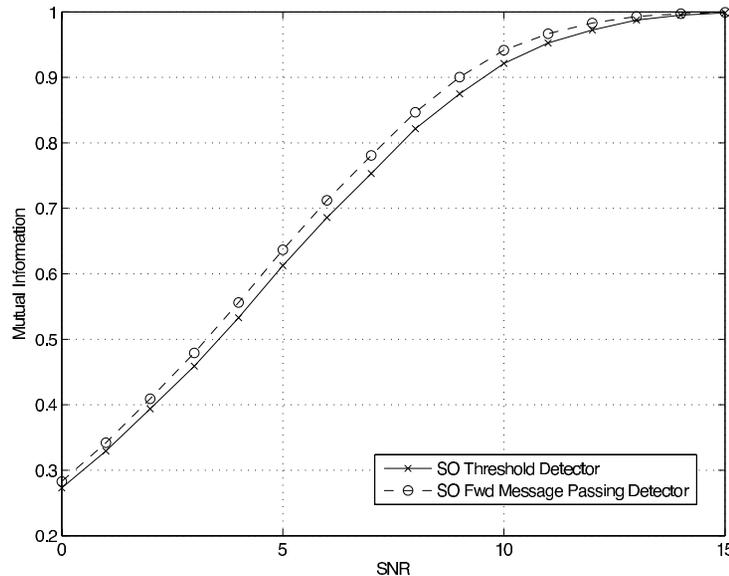
- $T_k$  is time-dependent. But, there are 4 time independent right null vectors and 2 time independent left null vectors.
- Forward recursion can be reduced to a  $3 \times 3$  recursion

# Reduced recursion.



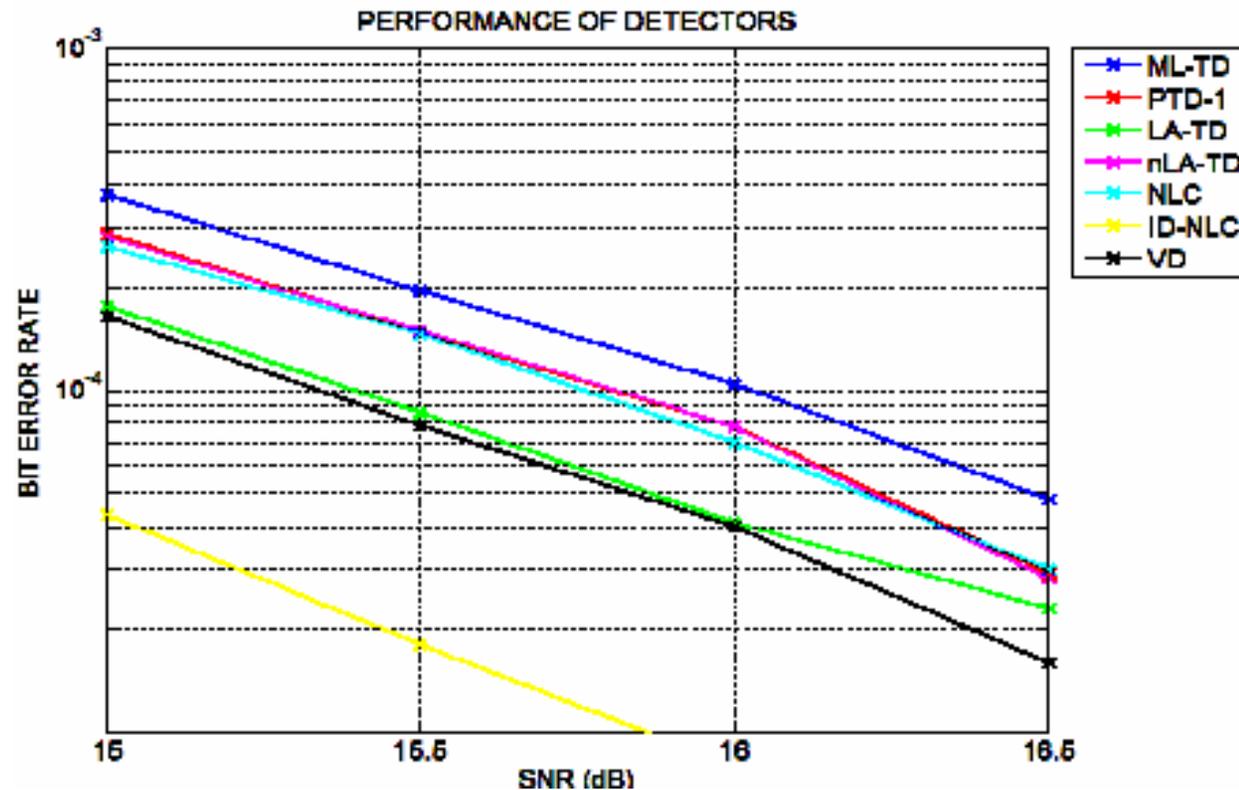
# Performance analysis

# Mutual information



- Mutual info between data and output LLR's:  
$$I(X, L) = \mathbf{E}_X \left( \int_{-\infty}^{\infty} dl \rho(l | x) \log_2 \left( \frac{\rho(l|x)}{\rho(l)} \right) \right)$$
- FMP detector resolves the asymmetry of LLR's.  
Channel capacity is increased by about 5% compared to THD channel

# Bit error rate



- The performance of FMP (green curve) matches the performance of Viterbi detector (black curve)

## Sector error rate: large deviations

**Outer code:**  $RS(w, \tau, nN)$ . **Inner code:** block size is  $nw + t$  bits. Symbol error counts for different IC blocks are independent identically distributed random variables. Let  $\vec{p} = \{p_0, p_1, \dots, p_n\}$  be the probability distribution of symbol error count  $\xi$  in an IC block such that  $\mathbb{E}(\xi) < \tau$ . Then

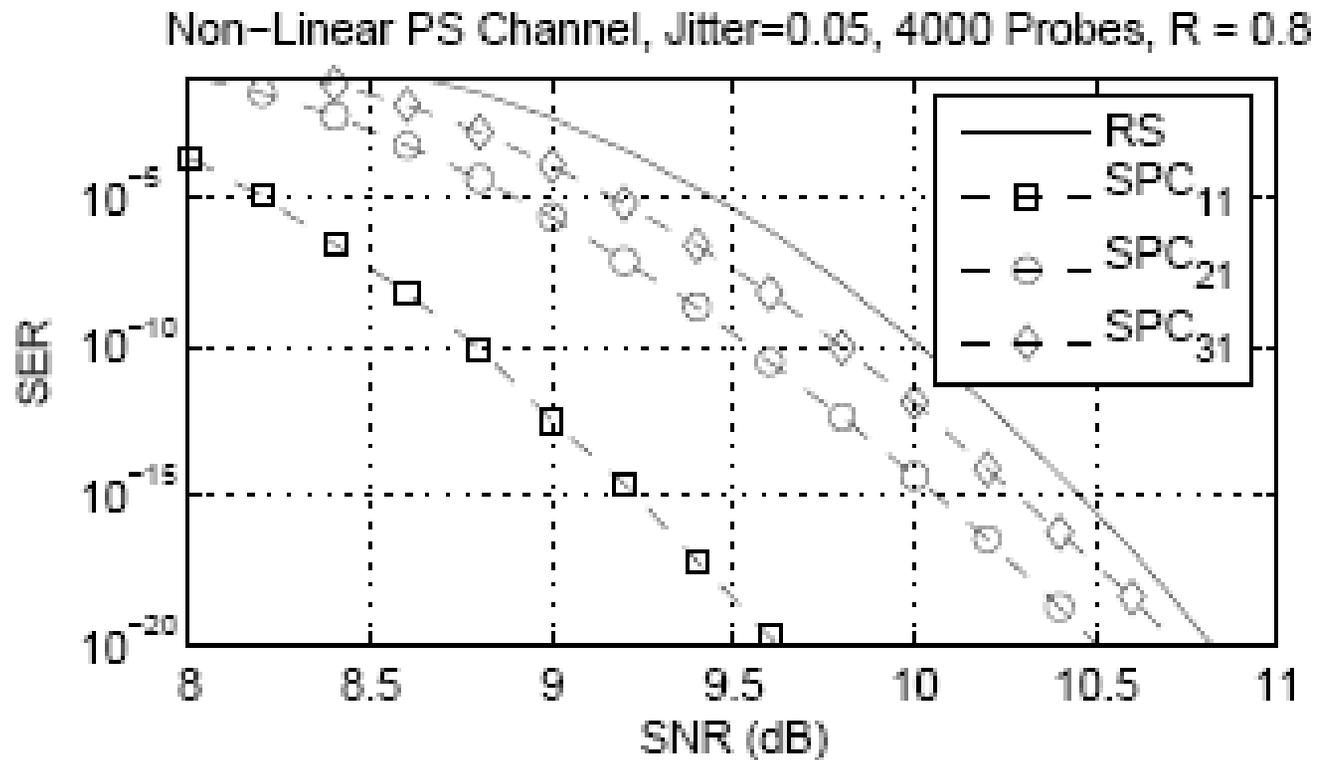
$$\frac{\ln(P_{se})}{N} \nearrow -D_{KL}(\vec{q}|\vec{p}), \text{ as } N \rightarrow \infty.$$

where  $\vec{q}$  is the effective probability distribution given by

$$q_k = \frac{p_k \mu^k}{\sum_{m=0}^n p_m \mu^m}, \quad k = 0, 1, \dots, n$$

and  $\mu$  is the unique positive solution of the critical point equation,  $\sum_{k=0}^n (k - n\tau) p_k \mu^k = 0$ ;  $D_{KL}$  is relative entropy

# Sector error rate comparison



## Conclusions-I

- Probe storage DSP is challenging to the extreme: on the one hand we have a very noisy channel, on the other - the allowed complexity of read channel per probe is severely restricted
- Forward message passing detector allows a generation of soft outputs at the complexity cost of a 3-state Viterbi detector without traceback unit with the performance matching that of the full 4-state Viterbi detector matched to the non-linear thermomechanical channel
- Large deviations analysis leads to an analytic expression for SER in probe storage, which is useful for sufficiently short inner codes
- Soft input  $SPC + RS$  code outperforms hard input  $RS$  code of the same rate by about 1 dB at  $SER = 10^{-15}$ .

## Conclusions-II

- A twist in the tale: asymptotically,  $RS$  code is better!
- Research supported by PROTEM FP6 European network grant
- The reported results will be published in the proceedings of ICC2008.